Ultrawideband Log-periodic Series-fed Printed Dipole Arrays Antenna

Chittawan Choeysakul, Sarawuth Chaimool, and Prayoot Akkaraekthalin
Electrical Engineering, Faculty of Engineering, King Mongkut’s Institute of Technology North Bangkok, Bangkok 10800, Thailand, E-mail: prayoot@kmitnb.ac.th

Abstract
By replacing a printed dipole of quasi-Yagi antenna by a series-fed printed dipole arrays with planar trapezoidal toothed log-periodic structure, a new ultrawideband operation and end-fire radiation pattern are presented. The experiment results of the proposed antenna, determined from 10-dB return loss, larger than 121% of the center frequency that covers the ultrawideband applications band and end-fire radiation pattern.

1. Introduction
The microstrip quasi-Yagi antenna exhibits attractive features including low profile, low-cost, lightweight, ease of fabrication, and broadband operation. These features make the microstrip quasi-Yagi antenna attractive for use in various applications. An end-fire quasi-Yagi antenna which is broadband uniplanar structure has been proposed for the first time by Kaneda et al. [1], achieving bandwidth 48% and since then explored in several papers[2]-[3]. Several bandwidth-enhancement techniques for printed dipoles or quasi-Yagi have been proposed in the literature, replacing printed dipole driver by printed bow-tie[4], patches driver [5] and a log-periodic dipole arrays director [6], tiled dipole [7] and several director elements [8]. However, those antennas have impedance bandwidths generally less than 85% of the center frequency.

In this paper, the printed dipole of quasi-Yagi antenna as driver and director is replaced by a series-fed printed dipole arrays with planar trapezoidal toothed log-periodic structure, which results in an improvement in bandwidth and gain. This new design combines a series-fed printed dipole arrays [9] and planar trapezoidal toothed log-periodic structure [10].

2. Antenna Design
Fig. 1 shows the ultrawideband log-periodic series-fed printed dipole arrays antenna. The top metallization consists of three parts, the microstrip-to-CPS transition, the series-fed printed-dipole arrays, and planar trapezoidal toothed log-periodic structure. The truncated microstrip ground plane on backside of the substrate is used as a reflector element for the antenna. We explain the three parts separately in order to understand the ultrawideband characteristics of the antenna. Optimizing those three parts separately makes the optimization easier.

2.1 Microstrip-to-CPS balun
Coplanar strip (CPS) is a uniplanar transmission line and a microstrip balun is usually desired to provide efficient transition between the CPS and the microstrip lines. The balun is used to introduce a half-wavelength

Figure 1: Schematic of the ultrawideband log-periodic series-fed printed dipole arrays antenna
delay, which is required to obtain the odd mode in the coupled microstrip line while suppressing the even mode
and establish a impedance matching between the balanced (dipole) and unbalanced (feed) elements of the antenna.
The balun is designed for the center frequency at 6 GHz, which creates 180° phase difference between the coupled
microstrip lines and offers a good impedance matching. Next, the length of driver ($L_{dri}$) and distance from the
driver to the reflector ($S_{ref}$) are sensitive elements. The antenna’s design frequency and its operational bandwidth
will be affected by these two parameters. Therefore, the length of the driver would be optimum when it is about
a guide-wavelength and the distance between the driver and the reflector is about a quarter guide wavelength. A
description of the microstrip-to-balun is provided in Fig. 2(a). A prototype has been designed using FR-4 substrate
($\varepsilon_r = 4.4$ and $\tan \delta = 0.022$). The thickness of the substrate is 1.6 mm. The optimization of the balun is also
done with the full-wave simulation. After a few optimization processes, the structural parameters of the balun are as
follows (unit: millimeters): $L_{dri}=38.6$, $L_1=7.208$, $L_2=5.424$, $L_3=12.124$, $L_4=7.332$, $L_5=3.5424$, $S_{ref}$ =11.2, $S_5=S_6
=0.6$, $W_{dri}=3.016$, $W_1=3.016$, $W_2=5.832$, $W_3=2.616$, $W_4=5.216$, $W_5=3.016$ and $W_6=1.608$.

2.2 Series-fed Printed Dipole Arrays

The modified quasi-Yagi antenna adopted new driver and director elements consists of changing the tradition
dipole director elements to a pair of series-fed printed dipole arrays in order to achieve maximum bandwidth. The
proposed series-fed printed dipole array is fed to the terminals of the longer element by using a direct connection
between the strip dipoles. This excitation method gives the phase progression for endfire radiation because the
CPS line does not support the even mode, which enables us to suppress the undesired mode excited in the coupled
microstrip line. We use a series-fed technique [7] that the operation bandwidth is restricted to the lengths of the
longest (for the lowest frequency) and the shortest (for the highest frequency) dipole elements of the antenna arrays.
The end of the series-fed is a parasitic director. Its dimensions are designed for the end of operation frequency.
The parasitic director element on the top plane simultaneously directs the antenna propagation toward the end-fire
direction, and acts as an impedance matching element. The dipoles with different lengths and director are printed
on the same side of the dielectric substrate and are connected directly through a straight CPS as shown in Fig.
2(b). The series-fed arrays’ dimensions are (unit:millimeter) $P_1=6.88$, $P_2=5.66$, $P_3=5.4$, $P_4=4.75$, $P_5=4.39$, $d_1
=3.0$, $d_2=2.71$, $d_3=2.52$, $d_4=2.51$, $d_5=2.71$, $L_1=21.4$, $L_2=17.8$, $L_3=16.2$, $L_4=14.2$, $L_5=12.2$, $L_{dri}=7.2$, $W_{e1}=1.50$,$W_{e2}=1.3$, $W_{e3}=1.2$, $W_{e4}=1.0$, $W_{e5}=0.95$, $W_{dri}=2.216$ and $S_{dir}=3.284$.

2.3 Planar Trapezoidal Toothed Log-periodic Structure

The antenna is based on the well-known log-periodic antenna scaling principle. Following this basis, the antenna
displays a periodic behaviour with frequency logarithm. The structure of the proposed antenna is shown in Fig.
2(c). The printed dipoles are periodically scaled with the longer element designed for the lower resonant
frequency and the shorter one for the higher frequency. The antenna dimensions follow the periodicty relationship :
if $f_1$ and $f_2$ are one period apart, they will be related to geometric ratio ($\tau$) by

$$\Delta = \ell n(f_2) - \ell n(f_1) = \ell n\left(\frac{1}{\tau}\right)$$

Figure 2: Schematic of the proposed antenna parts (a) microstrip-to-CPS, (b) the series-fed printed dipole arrays,
and (c) planar trapezoidal toothed log-periodic structure
Typical designs of planar trapezoidal toothed log-periodic dipole have apex half angles ($\alpha$) as

$$10^\circ \leq \alpha \leq 45^\circ$$

and

$$0.95 \geq \tau \geq 0.7$$

(2)

Maximum value of frequency span ($\Delta$) is related with minimum value of $\tau$, as 0.7. In Fig. 2(c), the dimensions of planar trapezoidal toothed log-periodic are defined by the inverse of the geometric ratio $\tau$

$$\frac{1}{\tau} = \frac{l_{n+1}}{l_n} = \frac{R_{n+1}}{R_n}$$

(3)

Designing of log-periodic part has been described in [10] and hence it is not repeated here. From Fig. 2(c), the dimensions after adjustment are $\tau = 0.7$, the value of $\beta$ is estimated as value of $W_{dr1}$ as 3.016 mm, and the other important value is $\alpha$. Since the value of it is much, it is possible that the value interfere with truncated ground plane. Therefore $30^\circ$ (9.834 mm) was used after adjustment.

3. Simulation and Measurement Results

By putting the three parts individually optimized together, we can achieve moderately good performance. However, further optimization is required in order to achieve even better performance of the whole transition. The proposed antenna is fabricated on a FR-4 substrate with a dielectric constant of 4.4, a thickness of 1.6 mm, and a loss tangent of 0.022. Photograph of the proposed antenna are shown in Fig. 3. Measurements were carried out using an HP8510C network analyzer. Fig. 4(a) plots both simulation and measurement results for input return loss. The simulated and measured bandwidths (return loss $\leq 10$ dB) that have consistent term are 118% and 121%, respectively that cover the ultrawideband band. Additionally, the gain of the antenna have been measured and its can be observed that the peak gains increase to more than 6 dBi at 7GHz and 11 GHz as shown in Fig 4(b).

The measured H-plane and E-plane radiation patterns are demonstrated in Fig. 5 and 6, respectively. For comparison, we selected the patterns at 5.0, 7.0, and 9.0 GHz, approximately corresponding to the lower end, center, and upper end frequencies of the operating band of the proposed antenna. The shape of the radiation patterns was found to be similar for all three frequencies. It can be observed that, as expected, the radiation patterns of higher frequencies are irregular because the balun does not have a $180^\circ$ phase difference to feed the printed dipole arrays.
4. Conclusion

The ultrawideband log-periodic series-fed printed dipole arrays antenna has been proposed. The simulated and measured results from the designed antenna have been obtained. This new structure exhibits about 121% impedance bandwidth that covers the ultrawideband applications band. The radiation characteristics of proposed antenna have also measured at the selected frequencies, which represent the end-fire radiation pattern with a gain of about 3-7 dBi. Because of its ultrawideband operation, low cost, easy fabrication, and good radiation patterns, the proposed antenna should be very useful in the ultrawideband communication systems.

References